Identification (notes chapters 6-7, slides Identification_EMA.pdf) 1. Identification demo

- Inverse problems : model forms, data (s 7.1) Model forms for identification, discussion of residual terms Frequency domain least-squares solution, implicit NL Evaluation of results (ch 8)
- 1.2.3.4.5.





MS2SC PROVIR http://savoir.ensam.eu/moodle/course/view.php?id=1874 http://savoir.ensam.eu/moodle/course/view.php?id=490

Experimental modal analysis : data





Some standard texts for the vibration community

- [1] D. J. Ewins, Modal Testing: Theory and Practice. John Wiley and Sons, Inc., New York, NY, 1984.
- [2] W. Heylen et P. Sas, Modal analysis theory and testing. Katholieke Universteit Leuven, Departement Werktuigkunde, 2006
- [3] K. G. McConnell, Vibration Testing. Theory and Practice. Wiley Interscience, New-York, 1995

Experimental modal analysis: an inverse problem



LTI model forms

- See MATLAB control or scipy.signal.lti
- ss state-space $\begin{cases} \dot{x} \} = [A]_{N \times N} \{x\} + [B] \{u\}_{NA \times 1} \\ \{y\}_{NS \times 1} = [c] \{x\} + D \{u\} \end{cases}$
- Mechanical SS $\begin{cases} \dot{p} \\ \ddot{p} \end{cases} = \begin{bmatrix} 0 & I \\ -\left[\backslash \omega_{j}^{2} \\ \gamma \end{bmatrix} \Gamma \end{bmatrix} \begin{cases} p \\ \dot{p} \end{cases} + \begin{bmatrix} 0 \\ \phi_{j}^{T} b \end{bmatrix} \{u(t)\}$ $\{y(t)\} = \begin{bmatrix} c\phi_{j} & 0 \end{bmatrix} \begin{cases} p \\ \dot{p} \end{cases}$

• **tf**
$$H = \frac{\sum_{i} [b_{i}]_{NS \times NA} s^{i}}{\sum_{j} a_{j} s^{j}}$$

$$\begin{cases} \dot{q} \\ \ddot{q} \end{cases} = \begin{bmatrix} 0 & I \\ -M^{-1}K & -M^{-1}C \end{bmatrix} \begin{cases} q \\ \dot{q} \end{cases} + \begin{bmatrix} 0 \\ M^{-1}b \end{bmatrix} \{u(t)\} = \begin{bmatrix} c & 0 \end{bmatrix} \begin{cases} q \\ \dot{q} \end{cases}$$

- zpk (zero pole gain) $H = k \frac{\prod_{k < N} (s [z_j]_{NS \times NA})}{\prod_{j=1}^{N} (s p_j)}$
- residue (partial fraction expansion = modal form) $[H(s)]_{NS \times NA} = \sum_{j=1}^{N/2} \left(\frac{[R_j]_{NS \times NA}}{s - \lambda_j} + \frac{[\bar{R}_j]}{s - \bar{\lambda}_j} \right)$

Inverse problem



Objective function

- Least squares
 - Objective : minimize distance
 - Variants : Log-least squares, output-error, ...
- Singular value decomposition
 - Objective : constrained rank approximation

Circle fit / -3 dB

Assumed model form

$$[H(s)] \approx \frac{A_j}{s^2 + 2\zeta_j \omega_j s + \omega_j^2}$$

 Geometrical properties of Nyquist give ω,ζ

$$\zeta_j \approx \frac{\omega_b - \omega_a}{2\omega_n}$$

 Poor accounting of other contributions









(e)



This is done on the blackboard 8

-3dB failures



This is done on the blackboard 9

Example : FD polynomial

- This is done on the blackboard
- Section 7.4.1 (frequency domain ARX)



Classical linear system Id

💰 ld - feplot(3,'cax1')		– 🗆 X
File Feplot Edit View Debug Desktop Window He	p	۲ د
🖽 🖬 🗕 + 🖳 🖂 🗑 🕉 출 휴 홈 중 중 🤞	そ × 💡 団 国 ½ 😳 🎟	
idcom(2) properties 💥	iiplot(2) Test 🗶 Figure 6 🗶 Figure 7 🗶 Figure 8: StabD 🗶	
Stack X Ident X Unv X Channel X StabD X		
🕑 Generate Run		* * * * * *
Display Display		* * * * * *
Ftol 0.1	95 - ** * * * * * * * * * * * * *	* * * * * *
Dtol 10.0		
AutoldMain Renew		8 8 8 8 8
DispMode StabDiag Only		* * * * * *
CurPole 0.0		* * * * * *
CurLocal Estimate		* * * * * *
	80 75 70 65 60 55 500 1000 1500 2000 2500 3000 Frequency [Hz]	3500

Identification phases

- Initialization pick from :
 - Single pole estimate [1]
 - Stabilization diagram [2]
- Estimate by band (why ? [1,3])

- How can problems be detected ? [3-4]
- Re-optimize poles (why ?)



- [1] E. Balmes, « Frequency domain identification of structural dynamics using the pole/residue parametrization », IMAC 1996
- [2] P. Verboven, « Frequency-domain system identification for modal analysis », Ph.D. thesis, 2002.
- [3] G. Martin, « Méthodes de corrélation calcul/essai pour l'analyse du crissement », Ph.D. CIFRE SDTools, Arts et Metiers ParisTech, Paris, 2017
- [4] G. Martin, E. Balmes, et T. Chancelier, « Characterization of identification errors and uses in localization of poor modal correlation », MSSP 2017.

Stabilization diagrams



Stabilization diagrams & bias



Noise induced bias -> work by band



[1]

[2]

Inconsistence between data and family of models induces bias /variability

M. Böswald, « Analysis of the bias in modal parameters obtained with frequency-domain rational fraction polynomial estimators », Proceedings of ISMA 2016 El-kafafy, De Troyer, Peeters, Guillaume, « Fast maximum-likelihood identification of modal parameters with uncertainty intervals: A modal model-based formulation », MSSP, 2013

Low frequency noise bias



Effects of residual terms



Here : ignoring two nearby modes induces bias

In general : account for flexibility and inertia effects (lower/upper residual)

Pole/Residue with high and low residual terms



Pole/Residue with high and low residual terms



 Pole/Residue with high and low residual terms



Pole/Residue with high and low residual terms



 Pole/Residue with high and low residual terms



Identification strategies



Determination of poles : main differentiator between methods Two main strategies

- select in over-specified set
- build gradually

Frequency domain output error

$$\min_{\lambda_j} \min_{R_j, E, F} \left| [\alpha(s)]_{test} - \sum_{j \in \text{identified}} \left(\frac{[R_j]}{s - \lambda_j} + \frac{\left[\bar{R}_j \right]}{s - \bar{\lambda}_j} \right) + [E(s)] \right|^2$$

- + R_j,E,F implicit functions of λ_j
- $\frac{\partial J}{\partial \lambda_j} = \frac{\partial R^T R}{\partial \lambda_j} = 2R^T \frac{\partial R}{\partial \lambda_j}$ can be computed analytically
- around a pole gradient not very affected by error on other poles

SDT implements

- Fast optimization
- All λ_j changed at each step using gradient only



Balmes, PhD MIT 93 & http://www.sdtools.com/pdf/IMAC96id.pdf

Validations of results

- FRF Superposition
- Modeshape visualization
- Residue properties





Quality : an error criterion

Is the model well identified?

- Superpose measured and identified FRF for each sensor c, around each mode j
- Compute error on Nyquist





Test/identification error



For each sensor, each mode may have strong error/noise & low contribution Test error per mode/sensor

[1] G. Martin, E. Balmes, et T. Chancelier, « Characterization of identification errors and uses in localization of poor modal correlation », MSSP 2017.

Evaluate quality of identification

📣 ld - iiplot(2) ldFrf Edit View

File

X

30H 30H 913 (K \$ S 11 J.- 8:-+

Window

Help

-

Ilplot Insert Debug Desktop





Error analysis usage



Sorting needed to decide when to do nothing

Other errors : NL/LPV ≠ one system

- Non linear system : resonance dependent on input point / accurate positioning
- MMIF or identification per impact location shows significant dispersion
- Multiple identification results are not perfectly coherent







Other errors : local modes

• « Same mode » multiple times





- non structural masses generate global mode duplicates
- « small mass » ≠ « tiny peak » (if this were true proof mass dampers would not work)





Subspace methods : IO (ERA)

$$[h(k\Delta t)]_{pq} = \begin{bmatrix} [y(k)] & [y(k+j_1)] & \dots & [y(k+j_{q-1})] \\ [y(k+i_1)] & [y(k+i_1+j_1)] & \dots & [y(k+i_1+j_{q-1})] \\ \vdots & \vdots & \ddots & \vdots \\ [y(k+i_{p-1})] & [y(k+i_{p-1}+j_1)] & \dots & [y(k+i_{p-1}+j_{q-1})] \end{bmatrix}$$

$$[h(k\Delta t)]_{pq} = [C]_p [A]^{k-1} [B]_q = \begin{bmatrix} [C] \\ \vdots \\ [C] [A]^{j_{p-1}} \end{bmatrix} [A]^{k-1} [B] \dots & [A]^{j_{q-1}} [B] \end{bmatrix}$$

$$[h(k\Delta t)]_{pq} \approx [U] [\backslash S \backslash] [V]^T \qquad [A] = \left[[\backslash S \backslash]^{-1/2} [U]^T [h(2\Delta t)]_{pq} [V] [\backslash S \backslash]^{-1/2} \right]$$

$$[B] = [\backslash S \backslash]^{1/2} [V]^T \begin{bmatrix} [\Lambda] \\ [0] \\ [0] \end{bmatrix}_{qNA \times NA}$$

$$[C] = [[\backslash I \backslash] [0] [0]]_{NS \times pNS} [U] [\backslash S \backslash]^{1/2}$$

[1] J. N. Juang and R. S. Pappa, "An Eigensystem Realization Algorithm (ERA) for Modal Parameter Estimation and Model Reduction," *J. Guidance, Control, and Dynamics*, vol. 8, no. 5, pp. 620–627, 1985.



[2] E. Reynders, "System identification methods for (operational) modal analysis: review and comparison," Archives of Computational Methods in Engineering, vol. 19, no. 1, pp. 51–124, 2012.

Order selection : MMIF

- Mode indicator functions
- Related to force appropriation (do by hand)
- Appropriation excites a single mode (allow picture taking)



Order selection : MMIF

- Multi exciter allows spatial filter $\{u_i(t)\}_{NA} = \{w_i\}_{NS}a(t)$
- Appropriation : choose $\{w_i\}_{NS}$ to make other modes unobservable
- Keywords

Modal coordinates Modal filter



Order selection : SVD

Rank of H is limited

- Perfect for simulation
- Problem for real data
- More useful for data volume reduction

P. Guillaume & al. 2005 Improved techniques for outlier detection



Properties of residues

- $\frac{R_j}{s \lambda_j} = \frac{\left\{c\psi_j\right\} \left\{\psi_j^T b\right\} + \left\{c\psi_{j+1}\right\} \left\{\psi_{j+1}^T b\right\}}{s \lambda_j} \\ \left(\left[c_{\text{col}}\right] \left\{\psi_j\right\}\right)^T = \left\{\psi_j\right\}^T \left[b_{\text{col}}\right]$ Minimal
- Reciprocal
- Second order model (modal damping)

$$\frac{[R_j]}{s-\lambda_j} + \frac{\left[\bar{R}_j\right]}{s-\bar{\lambda}_j} = \frac{2(s+\zeta_j\omega_j)Re(R_j) + 2\omega_j\sqrt{1-\zeta_j^2}Im(R_j)}{s^2 + 2\zeta_j\omega_j s + \omega_j^2}$$
$$[R_j] = \frac{T_j}{i\omega_j\sqrt{1-\zeta_j^2}} = [c]\frac{\phi_j}{\sqrt{i\omega_j\sqrt{1-\zeta_j^2}}}\frac{\phi_j^T}{\sqrt{i\omega_j\sqrt{1-\zeta_j^2}}}[b]$$

- Modal mass (reciprocal, second order) $\frac{1}{\mu_j(c_{col})} = c_{col}\phi_j\phi_j^T b_{col} = (c_{col}\phi_j)^2 = T_{jcol} = i\omega_j\sqrt{1-\zeta_j^2}R_{jcol}$
- Second order non proportional damping